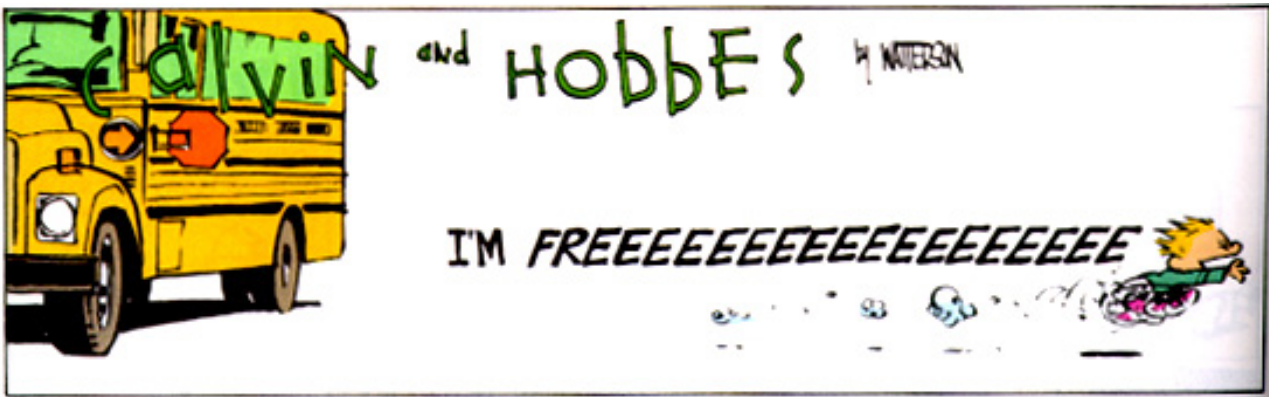


Name _____
ME161 Final. Wednesday December 8, 2004 7:00-10:00 p.m.

I certify that I upheld the Stanford Honor code during this exam _____



- Print your name and sign the honor code statement
- You may use your course notes, homework, books, etc.
- Write your answers on this handout
- Where space is provided, **show your work to get credit**
- If necessary, attach extra pages for scratch work
- Best wishes for a fun vacation. Merry Christmas and Happy New Year!

Problem	Value	Score
1	6	
2	13	
3	31	
4	20	
5	6	
6	15	
7	9	
Total	100	

Final.1 (6 pts.) Miscellaneous

(a) (1 pt.) **Describe one or two things you enjoyed learning in ME161.**

Result:

(b) (1 pt.) Name the engineer, scientist, or mathematician who first did the following:

Leonard Euler

- Considered sine and cosine as functions (not just the ratio of two sides of a triangle)
- Invented the symbols $\pi = 3.1415\dots$, $e = 2.71\dots$, $i = \sqrt{-1}$, \sum for summations, Δx for finite differences, and $f(t)$ for functions
- Discovered a general equation for rotational motions of a rigid body
- Created the foundations of the theory of differential equations
- Discovered a simple technique for numerically integrating differential equations
- Investigated analytical functions of a complex variable
- Invented the Taylor series expansion of a function (concurrent with MacLaurin)
- Studied mathematics under Johann Bernoulli at 14 years old
- Extended Bernoulli's equation for incompressible fluid flow
- Optimized the arrangement of masts on a ship
- "Wrote the book" on the mathematical theory of music
- Worked on cartography, magnetism, fire engines, ship building, and insurance
- Lost 8 of his 13 children in infancy, went half blind at 28 and mostly blind at 59, lost his home to a fire, and still had a positive attitude

(c) Any real, imaginary, or complex number can be expressed in magnitude-phase form.

- (1 pt.) Express -2 in magnitude-phase form.

$$-2 = 2 e^{(\pi + 2n\pi) i} \quad n=0, 1, 2, \dots$$

- (1 pt.) Why does multiplying two negative numbers produce a positive number? Using magnitude-phase form, show $-2 * -2 = +4$.

$$[2 e^{(\pi + 2n\pi) i}] * [2 e^{(\pi + 2n\pi) i}] = 4 e^{(2\pi + 4n\pi) i} = \cos(2\pi + 4n\pi) + i \sin(2\pi + 4n\pi) = 4$$

- (1 pt.) Complex numbers and exponentiation

Find *all* complex numbers (*in Cartesian form*) equal to the following.

$$\begin{aligned} \sqrt{i} &= \left[e^{(\frac{\pi}{2} + 2n\pi) i} \right]^{0.5} = e^{(\frac{\pi}{4} + n\pi) i} = \cos\left(\frac{\pi}{4} + n\pi\right) + i \sin\left(\frac{\pi}{4} + n\pi\right) \quad n=0, 1, 2, \dots \\ &= 0.707 + 0.707 i \quad \text{or} \quad -0.707 + -0.707 i \end{aligned}$$

- (1 pt.) Complex numbers and exponentiation

Find *all* complex numbers (*in Cartesian form*) equal to the following.

$$\begin{aligned} 1^{\frac{1}{2\pi}} &= (1 e^{2n\pi i})^{1/(2\pi)} = e^{n i} = \cos(n) + i \sin(n) \quad n=0, 1, 2, \dots \\ &= 1, \quad 0.54 + 0.84 i, \quad -0.42 + 0.91 i \quad \dots \end{aligned}$$

Final.2 (13 pts.) Sinusoidal transfer function and underdamped vibrations

A dynamic system's response is governed by $\ddot{y} + \dot{y} + y = f(t)$ where $f(t) = 10 \sin(2t)$.

- (a) (2 pts.) Find the system's transfer function

$$G(s) \triangleq \frac{Y(s)}{F(s)} = \frac{1}{s^2 + s + 1}$$

- (b) (3 pts.) Is the transfer function stable? **Yes**/No.

Explain: $G(s)$ is stable since its poles are in the left-half plane $(-0.5 \pm \sqrt{3}i)$.

- (c) (5 pts.) Find numerical values for the magnitude and phase of the sinusoidal transfer function.

$$\text{Magnitude} = \frac{1}{\sqrt{13}} = 0.27735$$

$$\text{Phase} = \text{atan2}(-2, -3) = -2.5536 \text{ rad}$$

- (d) (3 pts.) Find the steady-state response $y_{ss}(t)$.

$$y_{ss}(t) = 2.7735 * \sin(2t + -2.5536)$$

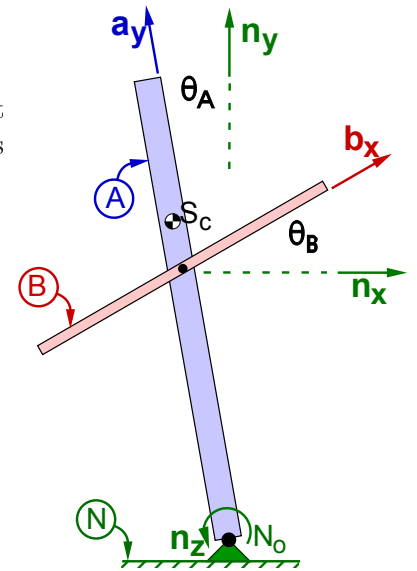
Final.3 (31 pts.) Solution and stability of a tight-rope walker

A tightrope walker A uses a rigid pole B to balance on a wire at a point N_o that is fixed in a Newtonian reference frame N . Right-handed sets of orthogonal unit vectors are fixed in N , A , and B , with:

- \mathbf{n}_x horizontal and to the right
- \mathbf{n}_y vertically upward
- \mathbf{a}_y directed from N_o to S_c (the mass center of A and B)
- \mathbf{b}_x directed along the balancing pole
- $\mathbf{n}_z = \mathbf{a}_z = \mathbf{b}_z$ perpendicular to the plane in which A and B move

The following identifiers are useful in this analysis.

Quantity	Symbol	Type
Mass of system formed by A and B	m	constant
Distance from N_o to S_c	d	constant
Related to mass distribution of system	I	constant
Central moment of inertia of B for \mathbf{b}_z	I^B	constant
Earth's sea-level gravitational constant	g	constant
Feedback-control torque on B from A	T_z	specified variable
Angle between \mathbf{n}_y and \mathbf{a}_y	θ_A	dependent variable
Angle between \mathbf{n}_x and \mathbf{b}_x	θ_B	dependent variable
Time	t	independent variable



(a) (1 pt.) The **nonlinear** equations of motion for the tightrope walker are

$$\begin{aligned} I \ddot{\theta}_A + -m g d \sin(\theta_A) &= -T_z \\ I^B \ddot{\theta}_B &= T_z \end{aligned}$$

Determine M , K , and G so that the **linearized** ODEs can be written in the matrix form $M \ddot{X} + K X = G T_z$ where $X \triangleq \begin{bmatrix} \theta_A \\ \theta_B \end{bmatrix}$.

Result:

$$\begin{bmatrix} I & 0 \\ 0 & I^B \end{bmatrix} \begin{bmatrix} \ddot{\theta}_A \\ \ddot{\theta}_B \end{bmatrix} + \begin{bmatrix} -m g d & 0 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} \theta_A \\ \theta_B \end{bmatrix} = \begin{bmatrix} -1 \\ 1 \end{bmatrix} T_z$$

(b) (3 pts.) Consider the homogeneous problem $T_z=0$. Assume a solution $X(t)=Ue^{pt}$ where p is a constant (to-be-determined) and U is a *non-zero* 2×1 matrix of constants (to-be-determined). Substitute this assumed solution into the matrix equation and find the equation that governs p and U . Express your results in terms of $\lambda \triangleq -p^2$, $A \triangleq M^{-1} K$, and the 2×2 identity matrix I .

Result:

$$(-\lambda I + A) U = 0$$

- (c) (4 pts.) Calculate the 2×2 matrix $A \triangleq M^{-1}K$ in terms of m, g, d, I^B and I .

Result:

$$A = \begin{bmatrix} \frac{-mgd}{I} & 0 \\ 0 & 0 \end{bmatrix}$$

- (d) (4 pts.) Calculate λ_i ($i = 1, 2$).

Result:

$$\lambda_1 = 0 \qquad \lambda_2 = \frac{-mgd}{I}$$

- (e) (2 pts.) Find p_1, p_2, p_3 and p_4 .

Result:

$$\begin{aligned} p_1 &= 0 & p_3 &= +\sqrt{\frac{mgd}{I}} \\ p_2 &= 0 & p_4 &= -\sqrt{\frac{mgd}{I}} \end{aligned}$$

- (f) (2 pts.) The solution for $X(t) \triangleq \begin{bmatrix} \theta_A(t) \\ \theta_B(t) \end{bmatrix}$ is **stable/neutrally stable/unstable**.

- (g) (2 pts.) A larger value of m corresponds to a **less/more** stable solution.

A larger value of I corresponds to a **less/more** stable solution.

A negative value of d corresponds to a **stable/neutrally stable/unstable** solution.⁴⁰

⁴⁰There are tightrope toys built with long downward curved balancing poles so the mass center of S is below N_o .

- (h) (6 pts.) Find U_1 and U_2 , the eigenvectors that correspond to λ_1 and λ_2 .

Result:

$$U_1 = \begin{bmatrix} 0 \\ 1 \end{bmatrix} \quad U_2 = \begin{bmatrix} 1 \\ 0 \end{bmatrix}$$

- (i) (2 pts.) Assemble the solution for $X(t)$ ⁴¹

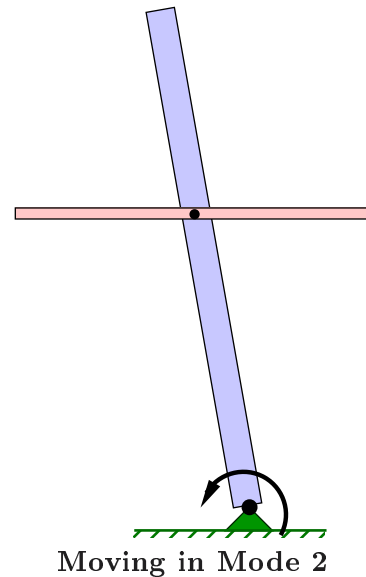
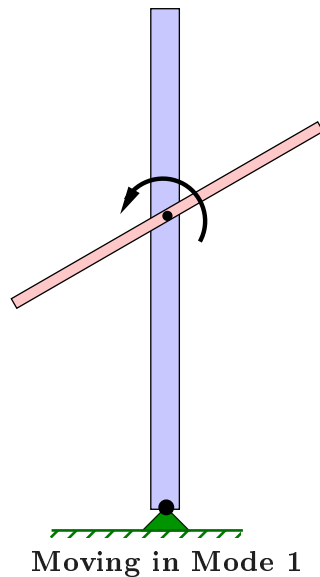
Result:

$$\begin{bmatrix} \theta_A(t) \\ \theta_B(t) \end{bmatrix} = c_1 \begin{bmatrix} 0 \\ 1 \end{bmatrix} + c_2 \begin{bmatrix} 0 \\ 1 \end{bmatrix} t + c_3 \begin{bmatrix} 1 \\ 0 \end{bmatrix} e^{+\sqrt{\frac{mgd}{I}} t} + c_4 \begin{bmatrix} 1 \\ 0 \end{bmatrix} e^{-\sqrt{\frac{mgd}{I}} t}$$

- (j) (1 pt.) How are the constants c_1, c_2, c_3, c_4 usually determined?

From the initial values of $\theta_A, \theta_B, \dot{\theta}_A,$ and $\dot{\theta}_B$.

- (k) (4 pts.) Sketch the system moving in its two modes. Clearly show how θ_A and θ_B are changing.

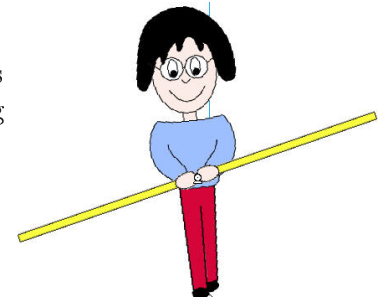


⁴¹When $T_2=0$, the original set of linear ODEs are uncoupled and one can find the homogeneous solution *without* matrices.

Final.4 (20 pts.) State-space feedback control of a tight-rope walker

One way to design an automatic control system for the tight-rope walker is to use the state-space method. The state space-method begins by defining the state matrix Y as

$$Y \triangleq \begin{bmatrix} \theta_A \\ \theta_B \\ \dot{\theta}_A \\ \dot{\theta}_B \end{bmatrix}$$



- (a) (1 pt.) Suppose the linearized ODEs for the tight-rope walker are

$$\begin{aligned} 100 \ddot{\theta}_A + -500 \theta_A &= -T_z \\ 25 \ddot{\theta}_B &= T_z \end{aligned}$$

Solve for $\ddot{\theta}_A$ and $\ddot{\theta}_B$ in terms of θ_A , θ_B , T_z , etc.,

Result:

$$\begin{aligned} \ddot{\theta}_A &= 5 \theta_A + -0.01 T_z \\ \ddot{\theta}_B &= 0.04 T_z \end{aligned}$$

- (b) (4 pts.) Cast these ODEs into the state-space form $\dot{Y} = AY + B_c T_z$.

Result:

$$\begin{bmatrix} \dot{\theta}_A \\ \dot{\theta}_B \\ \ddot{\theta}_A \\ \ddot{\theta}_B \end{bmatrix} = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ 5 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} \theta_A \\ \theta_B \\ \dot{\theta}_A \\ \dot{\theta}_B \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ -0.01 \\ 0.04 \end{bmatrix} T_z$$

- (c) (4 pts.) An engineer decides to balance the tight-rope walker using a feedback control law for T_z that is written in terms of the “feedback control constants” k_1, k_2, k_3, k_4 , as

$$T_z = k_1 \theta_A + k_2 \theta_B + k_3 \dot{\theta}_A + k_4 \dot{\theta}_B = [k_1 \quad k_2 \quad k_3 \quad k_4] \begin{bmatrix} \theta_A \\ \theta_B \\ \dot{\theta}_A \\ \dot{\theta}_B \end{bmatrix} = +K_c Y$$

Rewrite the ODEs in the state-space form $\dot{Y} = \bar{A}Y$ in terms of k_1, k_2, k_3, k_4 , and numbers.

Result:

$$\dot{Y} = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ 5 + -0.01 k_1 & -0.01 k_2 & -0.01 k_3 & -0.01 k_4 \\ 0.04 k_1 & 0.04 k_2 & 0.04 k_3 & 0.04 k_4 \end{bmatrix} Y$$

- (d) (1 pt.) Using physical intuition, **guess** at the signs of k_1 , k_2 , k_3 , and k_4 so that if you started with $\theta_A=10^\circ$ and $\theta_B=0^\circ$, you could bring this system to rest with $\theta_A=0^\circ$ and $\theta_B=0^\circ$. Circle $-$ if you think the number is negative, 0 if you think the number is zero, or $+$ if you think the number is positive. In other words, should the coefficients of θ_A , θ_B , $\dot{\theta}_A$, and $\dot{\theta}_B$ be negative, zero, or positive in order to bring this system to rest at $\theta_A = \theta_B = 0^\circ$.

Result:

$$T_z = (\underline{+ 0 -}) * \theta_A + (\underline{+ 0 -}) * \theta_B + (\underline{+ 0 -}) * \dot{\theta}_A + (\underline{+ 0 -}) * \dot{\theta}_B$$

- (e) (2 pts.) The process of solving for $Y(t)$ begins by assuming a solution of the form $Y(t) = U e^{\lambda t}$ where λ is a *constant* (to be determined), and U is a *non-zero* 4×1 matrix of constants (to be determined). After substituting this assumed solution into the governing ODE, the matrix equation that governs λ and U is

$$(-\lambda I + \bar{A}) U = 0$$

The unknowns in the previous equation are λ and U . Classify the previous matrix equation by picking the relevant qualifiers from the following list.

Uncoupled	Linear	Homogeneous	Algebraic
Coupled	Nonlinear	Inhomogeneous	Differential

- (f) (2 pts.) How does one solve for λ ?

Set the determinant of $(-\lambda I + \bar{A})$ equal to 0

- (g) (2 pts.) The polynomial equation that relates λ to k_1 , k_2 , k_3 , k_4 is (you do not need to show this)

$$\lambda^4 + (0.01 k_3 - 0.04 k_4) \lambda^3 + (0.01 k_1 - 0.04 k_2 - 5) \lambda^2 + (0.4 k_4) \lambda + (0.4 k_2) = 0$$

Determine whether k_i ($i=1,2,3,4$) must be **negative**, **zero**, **positive**, or **undetermined** for the roots of λ to have negative real parts so that $Y = U e^{\lambda t}$ is stable.

Result:

Feedback Control Constant	Negative, zero, positive, or undetermined
k_1	negative/zero/ <u>positive</u> /undetermined
k_2	negative/zero/ <u>positive</u> /undetermined
k_3	negative/zero/ <u>positive</u> /undetermined
k_4	negative/zero/ <u>positive</u> /undetermined

- (h) (1 pt.) Since the feedback control law is $T_z = k_1 \theta_A + k_2 \theta_B + k_3 \dot{\theta}_A + k_4 \dot{\theta}_B$ and you know some (or all) of the signs of k_1 , k_2 , k_3 , and k_4 , circle the correct signs in the following equation.

Result:

$$T_z = (\underline{+ 0 -}) * \theta_A + (\underline{+ 0 -}) * \theta_B + (\underline{+ 0 -}) * \dot{\theta}_A + (\underline{+ 0 -}) * \dot{\theta}_B$$

- (i) (2 pts.) Determine the values of λ that satisfy its polynomial equation when the system is uncontrolled, i.e., $k_1 = k_2 = k_3 = k_4 = 0$.

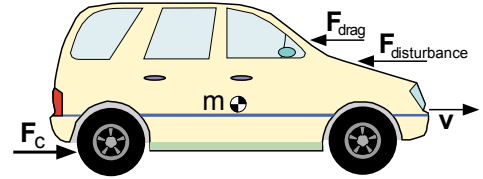
Result:

$$\begin{aligned} \lambda_1 &= 0 & \lambda_2 &= 0 \\ \lambda_3 &= +\sqrt{5} & \lambda_4 &= -\sqrt{5} \end{aligned}$$

- (j) (1 pts.) The uncontrolled system is **stable**/**neutrally stable**/**unstable**.

Final.5 (6 pts.) Cruise control for a car

A model for the speed v of a car of mass m includes unmodeled forces $F_{\text{disturbance}}$, an air-resistance drag force F_{Drag} , and the control force F_c (exerted by the engine and wheels) that tries to move the car at a desired (nominal) speed v_{nom} .



$$m \dot{v} = F_c - F_{\text{Drag}} - F_{\text{disturbance}}$$

- (a) **(2 pts.)** Determine $F_{c\text{nom}}$, the value of F_c required so $v = v_{\text{nom}}(t)$ when $F_{\text{disturbance}} = 0$.
 Note: v_{nom} is not necessarily constant.

$$F_{c\text{nom}} = m \dot{v}_{\text{nom}} + F_{\text{Drag}}$$

- (b) **(2 pts.)**

After separating F_c into two terms as shown to the right, rewrite the ODE in terms of \tilde{v} [the *error* between the actual value of v and the desired (nominal) value of v].

$$F_c = F_{c\text{nom}} + \tilde{F}_c$$

$$\tilde{v} \triangleq v - v_{\text{nom}}$$

$$m \dot{\tilde{v}} = \tilde{F}_c + F_{\text{disturbance}}$$

- (c) **(2 pts.)** What advantage does this model that includes air-resistance (drag) have over a model that does not include air-resistance? Explain.

The advantage of this improved model is that drag is eliminated from the disturbance forces, effectively reducing the magnitude of the disturbance force and improving the control system.

Final.6 (15 pts.) Proportional (P) feedback control for a simple car model

One choice for \tilde{F}_c is a *proportional control law* of the form $\tilde{F}_c = -k_p * \tilde{v}$ where k_p is a constant.

- (a) **(7 pts.)** Assuming $F_{\text{disturbance}}$ is *constant*, solve for $\tilde{v}(t)$ in terms of $F_{\text{disturbance}}$, m , k_p , t , and the initial error $\tilde{v}(0)$.

$$\tilde{v}(t) = \tilde{v}(0) e^{-\frac{k_p}{m} t} + \frac{-F_{\text{disturbance}}}{k_p} \left[1 - e^{-\frac{k_p}{m} t} \right]$$

- (b) (1 pt.) When $F_{\text{disturbance}}$ is constant, the steady-state error is $\tilde{v}_{\text{ss}} = \frac{-F_{\text{disturbance}}}{k_p}$
- (c) (1 pt.) Making k_p **small/large** & **negative/positive** gives the most stable solution for $\tilde{v}(t)$
- (d) (6 pts.) Find the values of k_p that satisfy the following specifications for a 1000 kg car.

Design specification	k_p
The steady-state error in response to $F_{\text{disturbance}} = 800 \text{ n}$ is $-2 \frac{\text{m}}{\text{sec}}$	400
With $F_{\text{disturbance}} = 0$, the car goes from rest to 90% of its desired speed within 5 sec	460.5

Final.7 (9 pts.) Proportional-Integral (PI) cruise control for a simple car model

One choice for \tilde{F}_c is a *proportional-integral control law* $\tilde{F}_c = -k_p * \tilde{v} + -k_i * \int_{\bar{t}=0}^t \tilde{v} d\bar{t}$ where k_p and k_i are constants.

- (a) (3 pts.) Write a 2^{nd} -order, inhomogeneous, ODE in standard form for $\tilde{v}(t)$.

$$m \ddot{\tilde{v}} + k_p \dot{\tilde{v}} + k_i \tilde{v} = \dot{F}_{\text{disturbance}}$$

- (b) (1 pt.) When $F_{\text{disturbance}}$ is constant, the steady-state error is $\tilde{v}_{\text{ss}} = 0$
- (c) (5 pts.) Find values of k_p and k_i that satisfy the following specifications for a 1000 kg car.

Design specification	k_p	k_i
With $F_{\text{disturbance}} = 0$, the car goes from rest and settles to within 1% of its desired speed within 5 sec with a maximum overshoot of 2%	2 Jblahblah	2 Jblahblah