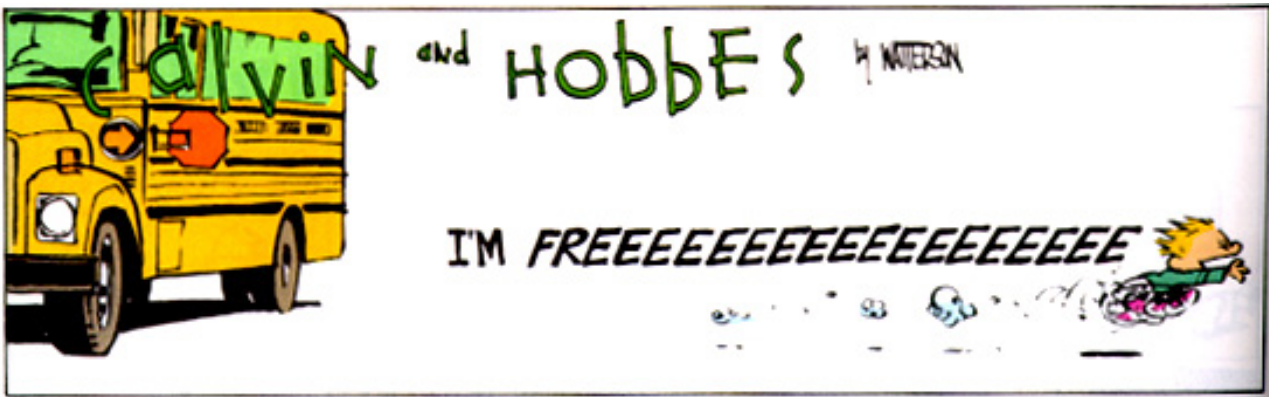


Name _____
ME161 Final. Wednesday December 8, 2004 7:00-10:00 p.m.

I certify that I upheld the Stanford Honor code during this exam _____



- Print your name and sign the honor code statement
- You may use your course notes, homework, books, etc.
- Write your answers on this handout
- Where space is provided, **show your work to get credit**
- If necessary, attach extra pages for scratch work
- Best wishes for a fun vacation. Merry Christmas and Happy New Year!

Problem	Value	Score
1	6	
2	13	
3	31	
4	20	
5	6	
6	15	
7	9	
Total	100	

Final.1 (6 pts.) Miscellaneous

(a) (1 pt.) **Describe one or two things you enjoyed learning in ME161.**

Result:

(b) (1 pt.) **Name the engineer, scientist, or mathematician who first did the following:**

- Considered sine and cosine as functions (not just the ratio of two sides of a triangle)
- Invented the symbols $\pi = 3.1415\dots$, $e = 2.71\dots$, $i = \sqrt{-1}$, \sum for summations, Δx for finite differences, and $f(t)$ for functions
- Discovered a general equation for rotational motions of a rigid body
- Created the foundations of the theory of differential equations
- Discovered a simple technique for numerically integrating differential equations
- Investigated analytical functions of a complex variable
- Invented the Taylor series expansion of a function (concurrent with MacLaurin)
- Studied mathematics under Johann Bernoulli at 14 years old
- Extended Bernoulli's equation for incompressible fluid flow
- Optimized the arrangement of masts on a ship
- "Wrote the book" on the mathematical theory of music
- Worked on cartography, magnetism, fire engines, ship building, and insurance
- Lost 8 of his 13 children in infancy, went half blind at 28 and mostly blind at 59, lost his home to a fire, and still had a positive attitude

(c) **Any real, imaginary, or complex number can be expressed in magnitude-phase form.**

- (1 pt.) Express -2 in magnitude-phase form.

$$-2 = e^{(\quad + 2n\pi)i} \quad n=0, 1, 2, \dots$$

- (1 pt.) **Why does multiplying two negative numbers produce a positive number?**
Using magnitude-phase form, show $-2 * -2 = +4$.

- (2 pts.) **Complex numbers and exponentiation**

Find **all** complex numbers (in Cartesian form) equal to the following.

$$\sqrt{i} = \underline{\quad} + \underline{\quad}i \quad \text{or} \quad \underline{\quad} + \underline{\quad}i$$

$$1^{\frac{1}{2\pi}} = 1,$$

Final.2 (13 pts.) Sinusoidal transfer function and underdamped vibrations

A dynamic system's response is governed by $\ddot{y} + \dot{y} + y = f(t)$ where $f(t) = 10 \sin(2t)$.

- (a) **(2 pts.)** Find the system's transfer function

$$G(s) \triangleq \frac{Y(s)}{F(s)} =$$

- (b) **(3 pts.)** Is the transfer function stable? **Yes/No.**

Explain:

- (c) **(5 pts.)** Find numerical values for the magnitude and phase of the sinusoidal transfer function.

Magnitude =

Phase =

rad

- (d) **(3 pts.)** Find the steady-state response $y_{ss}(t)$.

$$y_{ss}(t) =$$

- (c) **(4 pts.)** Calculate the 2×2 matrix $A \triangleq M^{-1}K$ in terms of m, g, d, I^B and I .

Result:

$$A = \begin{bmatrix} & \\ & \end{bmatrix}$$

- (d) **(4 pts.)** Calculate λ_i ($i = 1, 2$).

Result:

$$\lambda_1 = \qquad \qquad \lambda_2 =$$

- (e) **(2 pts.)** Find p_1, p_2, p_3 and p_4 .

Result:

$$p_1 = \qquad \qquad p_3 =$$

$$p_2 = \qquad \qquad p_4 =$$

- (f) **(2 pts.)** The solution for $X(t) \triangleq \begin{bmatrix} \theta_A(t) \\ \theta_B(t) \end{bmatrix}$ is **stable/ neutrally stable/ unstable**.

- (g) **(2 pts.)** A larger value of m corresponds to a **less/more** stable solution.
A larger value of I corresponds to a **less/more** stable solution.

- (h) **(6 pts.)** Find U_1 and U_2 , the eigenvectors that correspond to λ_1 and λ_2 .

Result:

$$U_1 = \begin{bmatrix} \\ \end{bmatrix} \qquad U_2 = \begin{bmatrix} \\ \end{bmatrix}$$

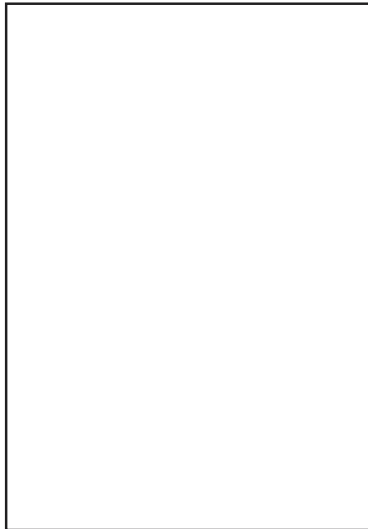
- (i) **(2 pts.)** Assemble the solution for $X(t)$

Result:

$$\begin{bmatrix} \theta_A(t) \\ \theta_B(t) \end{bmatrix} = c_1 \begin{bmatrix} \\ \end{bmatrix} + c_2 \begin{bmatrix} \\ \end{bmatrix} + c_3 \begin{bmatrix} \\ \end{bmatrix} + c_4 \begin{bmatrix} \\ \end{bmatrix}$$

- (j) **(1 pt.)** How are the constants c_1, c_2, c_3, c_4 usually determined?

- (k) **(4 pts.)** Sketch the system moving in its two modes. Clearly show how θ_A and θ_B are changing.



Moving in Mode 1

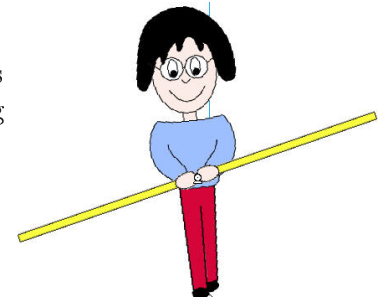


Moving in Mode 2

Final.4 (20 pts.) State-space feedback control of a tight-rope walker

One way to design an automatic control system for the tight-rope walker is to use the state-space method. The state space-method begins by defining the state matrix Y as

$$Y \triangleq \begin{bmatrix} \theta_A \\ \theta_B \\ \dot{\theta}_A \\ \dot{\theta}_B \end{bmatrix}$$



- (a) (1 pt.) Suppose the linearized ODEs for the tight-rope walker are

$$\begin{aligned} 100\ddot{\theta}_A + -500\theta_A &= -T_z \\ 25\ddot{\theta}_B &= T_z \end{aligned}$$

Solve for $\ddot{\theta}_A$ and $\ddot{\theta}_B$ in terms of θ_A , θ_B , T_z , etc.,

Result:

$$\begin{aligned} \ddot{\theta}_A &= \\ \ddot{\theta}_B &= \end{aligned}$$

- (b) (4 pts.) Cast these ODEs into the state-space form $\dot{Y} = AY + B_c T_z$ by completing the following matrices.

Result:

$$\begin{bmatrix} \dot{\theta}_A \\ \dot{\theta}_B \\ \ddot{\theta}_A \\ \ddot{\theta}_B \end{bmatrix} = \begin{bmatrix} & & & \\ & & & \\ & & & \\ & & & \end{bmatrix} \begin{bmatrix} \theta_A \\ \theta_B \\ \dot{\theta}_A \\ \dot{\theta}_B \end{bmatrix} + \begin{bmatrix} & & & \\ & & & \\ & & & \\ & & & \end{bmatrix} T_z$$

- (c) (4 pts.) An engineer decides to balance the tight-rope walker using a feedback control law for T_z that is written in terms of the “feedback control constants” k_1 , k_2 , k_3 , k_4 , as

$$T_z = k_1 \theta_A + k_2 \theta_B + k_3 \dot{\theta}_A + k_4 \dot{\theta}_B = \begin{bmatrix} k_1 & k_2 & k_3 & k_4 \end{bmatrix} \begin{bmatrix} \theta_A \\ \theta_B \\ \dot{\theta}_A \\ \dot{\theta}_B \end{bmatrix} = +K_c Y$$

Rewrite the ODEs in the state-space form $\dot{Y} = \bar{A}Y$ in terms of k_1 , k_2 , k_3 , k_4 , and numbers.

Result:

$$\dot{Y} = \begin{bmatrix} & & & \\ & & & \\ & & & \\ & & & \end{bmatrix} Y$$

- (d) (1 pt.) Using physical intuition, **guess** at the signs of k_1 , k_2 , k_3 , and k_4 so that if you started with $\theta_A=10^\circ$ and $\theta_B=0^\circ$, you could bring this system to rest with $\theta_A=0^\circ$ and $\theta_B=0^\circ$. Circle **-** if you think the number is negative, **0** if you think the number is zero, or **+** if you think the number is positive. In other words, should the coefficients of θ_A , θ_B , $\dot{\theta}_A$, and $\dot{\theta}_B$ be negative, zero, or positive in order to bring this system to rest at $\theta_A = \theta_B = 0^\circ$.

Result:

$$T_z = (\underline{+ \ 0 \ -}) * \theta_A + (\underline{+ \ 0 \ -}) * \theta_B + (\underline{+ \ 0 \ -}) * \dot{\theta}_A + (\underline{+ \ 0 \ -}) * \dot{\theta}_B$$

- (e) (2 pts.) The process of solving for $Y(t)$ begins by assuming a solution of the form $Y(t) = U e^{\lambda t}$ where λ is a *constant* (to be determined), and U is a *non-zero* 4×1 matrix of constants (to be determined). After substituting this assumed solution into the governing ODE, the matrix equation that governs λ and U is

$$(-\lambda I + \overline{A}) U = 0$$

The unknowns in the previous equation are λ and U . Classify the previous matrix equation by picking the relevant qualifiers from the following list.

Uncoupled	Linear	Homogeneous	Algebraic
Coupled	Nonlinear	Inhomogeneous	Differential

- (f) (2 pts.) How does one solve for λ ?
- (g) (3 pts.) The polynomial equation that relates λ to k_1 , k_2 , k_3 , and k_4 is (you do not need to show this)

$$\lambda^4 + (0.01 k_3 - 0.04 k_4) \lambda^3 + (0.01 k_1 - 0.04 k_2 - 5) \lambda^2 + (0.4 k_4) \lambda + (0.4 k_2) = 0$$

Determine whether k_i ($i=1,2,3,4$) must be **negative**, **zero**, **positive**, or **undetermined** for the roots of λ to have negative real parts so that $Y = U e^{\lambda t}$ is stable.

Result:

Feedback Control Constant	Negative, zero, positive, or undetermined
k_1	negative/zero/positive/undetermined
k_2	negative/zero/positive/undetermined
k_3	negative/zero/positive/undetermined
k_4	negative/zero/positive/undetermined

- (h) (1 pt.) Since the feedback control law is $T_z = k_1 \theta_A + k_2 \theta_B + k_3 \dot{\theta}_A + k_4 \dot{\theta}_B$ and you know some (or all) of the signs of k_1 , k_2 , k_3 , and k_4 , circle the correct signs in the following equation.

Result:

$$T_z = (\underline{+ \ 0 \ -}) * \theta_A + (\underline{+ \ 0 \ -}) * \theta_B + (\underline{+ \ 0 \ -}) * \dot{\theta}_A + (\underline{+ \ 0 \ -}) * \dot{\theta}_B$$

- (i) (2 pts.) Determine the values of λ that satisfy its polynomial equation when the system is uncontrolled, i.e., $k_1 = k_2 = k_3 = k_4 = 0$.

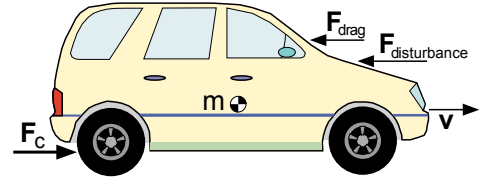
Result:

$$\begin{aligned} \lambda_1 &= & \lambda_2 &= \\ \lambda_3 &= & \lambda_4 &= \end{aligned}$$

- (j) (1 pts.) The uncontrolled system is **stable/ neutrally stable/ unstable**.

Final.5 (6 pts.) Cruise control for a car

A model for the speed v of a car of mass m includes unmodeled forces $F_{\text{disturbance}}$, an air-resistance drag force F_{Drag} , and the control force F_c (exerted by the engine and wheels) that tries to move the car at a desired (nominal) speed v_{nom} .



$$m \dot{v} = F_c - F_{\text{Drag}} - F_{\text{disturbance}}$$

- (a) (2 pts.) Determine $F_{c\text{nom}}$, the value of F_c required so $v = v_{\text{nom}}(t)$ when $F_{\text{disturbance}} = 0$.

$$F_{c\text{nom}} =$$

- (b) (2 pts.)

After separating F_c into two terms as shown to the right, rewrite the ODE in terms of \tilde{v} [the **error** between the actual value of v and the desired (nominal) value of v].

$$F_c = F_{c\text{nom}} + \tilde{F}_c$$

$$\tilde{v} \triangleq v - v_{\text{nom}}$$

=

- (c) (2 pts.) What advantage does this model that includes air-resistance (drag) have over a model that does not include air-resistance? Explain.

Final.6 (15 pts.) Proportional (P) feedback control for a simple car model

One choice for \tilde{F}_c is a **proportional control law** of the form $\tilde{F}_c = -k_p * \tilde{v}$ where k_p is a constant.

- (a) (7 pts.) Assuming $F_{\text{disturbance}}$ is **constant**, solve for $\tilde{v}(t)$ in terms of $F_{\text{disturbance}}$, m , k_p , t , and the initial error $\tilde{v}(0)$.

$$\tilde{v}(t) =$$

- (b) **(1 pt.)** When $F_{\text{disturbance}}$ is constant, the steady-state error is $\tilde{v}_{\text{ss}} =$
- (c) **(1 pt.)** Making k_p **small/large** & **negative/positive** gives the most stable solution for $\tilde{v}(t)$
- (d) **(6 pts.)** Find the values of k_p that satisfy the following specifications for a 1000 kg car.

Design specification	k_p
The steady-state error in response to $F_{\text{disturbance}} = 800 \text{ n}$ is $-2 \frac{\text{m}}{\text{sec}}$	
With $F_{\text{disturbance}} = 0$, the car goes from rest to 90% of its desired speed within 5 sec	

Final.7 (9 pts.) Proportional-Integral (PI) cruise control for a simple car model

One choice for \tilde{F}_c is a **proportional-integral control law** $\tilde{F}_c = -k_p * \tilde{v} + -k_i * \int_{\tilde{t}=0}^t \tilde{v} d\tilde{t}$ where k_p and k_i are constants.

- (a) **(3 pts.)** Write a 2^{nd} -order, inhomogeneous, ODE in standard form for $\tilde{v}(t)$.

=

- (b) **(1 pt.)** When $F_{\text{disturbance}}$ is constant, the steady-state error is $\tilde{v}_{\text{ss}} =$
- (c) **(5 pts.)** Find values of k_p and k_i that satisfy the following specifications for a 1000 kg car.

Design specification	k_p	k_i
With $F_{\text{disturbance}} = 0$, the car goes from rest and settles to within 1% of its desired speed within 5 sec with a maximum overshoot of 2%		